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Title: Estimation of Dynamic Interstory Drift in Buildings using Wireless Smart Sensors

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Abstract

Interstory drift response is one of most important quantities to quickly assess performance and damages in buildings. Nevertheless, direct measurement of interstory drift is difficult and expensive, because a stationary reference is required to attach measurement devices. With the goal of accurate and fast reference-free estimation, this paper proposes a new strategy to determine dynamic interstory drifts using accelerations. In particular, a Tikhonov regularization is adopted in a generalized minimization problem to achieve an efficient and stable FIR filter. Furthermore, due to independent clocks in wireless sensors, accurate time synchronization of the records is critical, and consequently, a strategy for accurate synchronization is also presented. Finally, the proposed strategy has been deployed on edge devices for onboard real-time interstory drift estimation. The proposed method for dynamic interstory drift estimation is validated, first, by numerical simulation using earthquake records as base excitation of linear and nonlinear buildings, as well as through laboratory shake table experiments. Both numerical and lab test results show good agreement of dynamic interstory drifts between the measured value and estimated results, demonstrating the efficacy of the proposed method to estimate the dynamic displacements of seismically excited structures.

KEYWORDS: Reference-free displacement estimation, smart wireless sensor, FIR filter, time synchronization.

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29

1. INTRODUCTION

30 Measurement of structural displacements under service or extreme loads are typically
31 desired for applications such as calibrating structural designs and assessing structural
32 performance (Skolnik and Wallace, 2010). Furthermore, buildings interstory drifts are
33 recognized as a critical quantity to estimate structural performance and damage (Bennett and
34 Batrone, 1997). Direct measurement of interstory drift is difficult because it is the difference
35 of displacement between two stories, e.g., using linear variable differential transformers and
36 laser-based sensors (Islam et al, 2016). Alternatively, displacement measurements can be
37 obtained directly by non-contact sensors from a remote location, e.g., using laser Doppler
38 vibrometers which can be quite expensive (Kim and Sohn, 2017). Recently, computer vision
39 systems, such as commercial grade cameras/smartphones or unmanned aerial systems has
40 received increasing attention. However, they usually require a reference on the video which
41 may not be readily available (Luo and Feng, 2018; Lee et al, 2020). In addition, the sampling
42 frequency is limited by the typically low camera frame rate, the visibility is governed by
43 environmental and lighting conditions, and the accuracy is affected by long distances from the
44 viewpoint to the region of interest. Note that the existing solutions are not well-suited for
45 interstory drift measurement of real-scale buildings as either the cost are prohibitive, a reference
46 is needed, or implementation is challenging.

47 Researchers have proposed to calculate interstory drifts indirectly from estimated
48 displacement using other measurements such as velocities, accelerations, and/or strains.
49 Estimation using accelerations has good potential because of the ease and low cost to measure
50 accelerations reliably. However, displacement estimation using double integration diverges,
51 because it amplifies the noise in the acceleration, especially in low-frequency domain. Many
52 solutions have been proposed for this problem (Kim et al, 2014; Nagayama et al, 2017; Hester
53 et al, 2017; Abé and Fujino, 2017; Gindy et al, 2008; Liu et al, 2017). For example, a recursive
54 high-pass filter and a recursive integrator are proposed to achieve real-time online displacement
55 estimation by means of multi-round baseline correction, filtering, and integration (Zheng et al,
56 2019). This method has yet to be implemented in edge devices for real-time demonstration. An
57 extended Kalman filter with an embedded Bayesian noise-parameter updating has also been
58 proposed to reduce numerical errors in displacement estimation from seismic accelerations.
59 However, it requires a nonlinear model, which may not be available for many scenarios (Pan et
60 al, 2021). To improve the accuracy, the author has proposed an approach to minimize the L2-
61 norm of a functional with a Tikhonov regularization, which represents a higher-order derivative
62 of the difference of the measured acceleration and the second derivative of the estimated
63 displacement (Gomez et al, 2018). On the other hand, the residual deformation, corresponding
64 to the DC component in the frequency domain, is not able to be captured by integrating. To
65 address this concern, many researchers consider data fusion, leveraging another type of sensors

66 which can capture the low-frequency component and stitching it together with the information
67 obtained from accelerometers (Park et al, 2013; Zhu et al, 2020; Park et al, 2018; Kim et al,
68 2018). This paper is focused on deployment scenarios where only accelerometers are available,
69 which is very common for full-scale deployment of wireless smart sensors in buildings. Indeed,
70 acceleration is the most reliable and popular measurement, mainly because accelerometers are
71 easy to install and do not require complex surface mounting.

72 In this paper, the goal is to estimate interstory drift from acceleration-only measurements
73 using wireless smart sensors, with a focus on time synchronization while comparing
74 displacement estimation from different sensors. Wireless smart sensors (WSS) are cost-
75 effective small-size integrated data acquisition devices, which consist of sensors (most often
76 accelerometers), computing unit, wireless transceiver, and/or actuation interface (Lynch et al,
77 2006; Rawat et al, 2014). Major efforts have been spent on developing WSS prototypes with
78 advancements both in hardware and software, e.g., iMote2 and Xnode developed by researchers
79 from University of Illinois (Rice et al, 2010; Rice et al, 2011; Jo et al, 2011; Spencer et al, 2017;
80 Fu et al, 2016; Fu et al, 2019). While efficient for displacement estimation, WSS have several
81 inherent challenges that must be addressed, one of which is time synchronization. In particular,
82 WSS use local clocks, which do not share a global time and they drift at different rates.
83 Furthermore, synchronization of local clocks of different sensor nodes does not guarantee the
84 synchronization of measurement data obtained from each sensor node (Nagayama and Spencer,

85 2007). The usage of unsynchronized data may negatively affect subsequent analysis, especially
86 for interstory drift estimation, which relies on the relative displacement estimations between
87 each pair of sensors. Though some studies have developed and implemented the technologies
88 of time synchronization on several WSS platforms (Wang et al, 2007; Kim et al, 2010; Bocca
89 et al, 2011), most of the work solely considers clock and not data synchronization.

90 This study proposes the use of a FIR filter via Tikhonov regularization to estimate accurate
91 dynamic interstory drifts in buildings based on acceleration measurements at different floors.
92 Furthermore, an efficient time synchronization strategy is proposed to enable the usage of
93 wireless smart sensors to obtain accurate dynamic interstory drift estimation. The filter method
94 together with the time synchronization strategy is finally deployed on a network of WSS and
95 executed onboard using limited computational resources.

96

97 2. DYNAMIC INTERSTORY DRIFT ESTIMATION FROM ACCELERATION RECORDS

98 The use of Tikhonov regularization to estimate dynamic displacements was first proposed
99 by Hong's group (Hong et al, 2010; Lee et al, 2010), and subsequently improved by Gomez et
100 al (2018) to estimate dynamic reference-free bridge displacements. Among other dynamic
101 displacement estimation algorithms, it gives the best accuracy, introduce zero phase delays in
102 the measurement, and calculates the results in an efficient time. Therefore, this idea is adopted
103 in this study in buildings to estimate dynamic interstory drifts. For the convenience of the reader,

104 a brief overview of the filter formulation is presented in the following subsection.

105 2.1 Displacement estimation formulation

106 The following functional with Tikhonov regularization represents the error in a high-order
107 derivative of the difference between estimated displacements and measured accelerations
108 (Gomez et al, 2019),

$$109 \quad \Pi(u) = \frac{1}{2} \int_T \left[\frac{d^{n-2}}{dt^{n-2}} \left(\frac{d^2 u}{dt^2} - \bar{a} \right) \right]^2 dt + \frac{1}{2} \beta^n \int_T u^2 dt \quad (1)$$

110 where T is the time window of interest $t_1 < t < t_2$, \bar{a} is the measured acceleration, u is
111 the estimated displacement, $n \geq 2$ is an integer named as the order of the functional, and
112 $\beta > 0$ is a parameter known as the factor of Tikhonov regularization.

113 From the previous functional, the following ordinary differential equation is obtained
114 based on the variational calculus.

$$115 \quad \frac{d^{2n} u}{dt^{2n}} + (-\beta)^n u = \frac{d^{2n-2} \bar{a}}{dt^{2n-2}}, \quad t_1 < t < t_2 \quad (2)$$

116 To achieve a unique solution, the boundary conditions (BC) for the differential equation are
117 known displacements and derivatives at the ends of the time window t_1, t_2 . However, these BC
118 are not available for this application. Although, the BC affect the solution close to ends of the
119 time window, their influence is smaller towards the center of the time window. Consequently,
120 using superposing moving windows centered at each time point minimize the effect of the
121 unknown BC.

122 Following the Fourier transform of the differential equation, the frequency response is
 123 obtained.

$$124 \quad H_{\bar{a}}(\omega) = -\frac{\omega^{2n-2}}{\omega^{2n} + \beta^n} \quad (3)$$

125 The regularization factor β is computed by defining a target frequency f_T and target accuracy
 126 α_T , obtaining the following expression.

$$127 \quad \beta = \sqrt[n]{\frac{1 - \alpha_T}{\alpha_T}} (2\pi f_T)^2 \quad (4)$$

128 The previous continuous-time representation is unstable, therefore, a discrete FIR filter is
 129 considered instead, a FIR filter of type I with generalized linear phase is chosen. The FIR filter
 130 is represented by a vector of coefficients \mathbf{c} with the length of $2k+1$. Consequently, the estimated
 131 displacement is given in terms of the measured acceleration and the FIR filter by the formula

$$132 \quad u(t) = (\Delta t)^2 \sum_{p=-k}^k c_{k+1+p} \bar{a}(t + p\Delta t) \quad (5)$$

133 where the coefficients of filter are given by

$$134 \quad c_{p+k+1} = -\frac{f_s}{2\pi^2} \int_0^{f_s/2} \frac{f^{2n-2}}{f^{2n} + \lambda^{2n} f_T^{2n}} \cos(2\pi p f \Delta t) df \quad (6)$$

135 The filter length is determined by the normalized window length N_w , the target frequency
 136 f_T , and the sampling frequency f_s , expressed as,

$$137 \quad k = N_w \frac{f_s}{2f_T} \quad (7)$$

138 The normalized time window is defined, so as to make the impulse response function end at

139 zeros in both ends, which minimizes Gibbs' phenomenon and rippling in the frequency domain.

140 Appropriate values for the normalized window length N_w are presented in (Gomez et al, 2019).

141 2.2 FIR filter for interstory drift estimation

142 To estimate dynamic interstory drifts, acceleration records two consecutive floors are
143 required: \bar{a}_i and \bar{a}_{i+1} . The measurements need to be synchronized in time and have the same
144 sampling frequency. For wired sensors, time synchronization is not a problem even for long
145 cables, but in wireless smart sensors, it is a major challenge to overcome. Section 3 describes
146 the details of this problem and a solution strategy.

147 The dynamic interstory drifts are estimated using the FIR filter proposed in the previous
148 section, which is defined by the vector of coefficients \mathbf{c} . Consequently, the estimated
149 displacements at the two locations are given in terms of the measured accelerations and the FIR
150 filter by the formula

$$151 \quad u_i(t) = (\Delta t)^2 \sum_{p=-k}^k c_{k+1+p} \bar{a}_i(t + p\Delta t) \quad (8)$$

$$152 \quad u_{i+1}(t) = (\Delta t)^2 \sum_{p=-k}^k c_{k+1+p} \bar{a}_{i+1}(t + p\Delta t) \quad (9)$$

153 where the coefficients of filter are given by Eq. (6) and the filter length is given by Eq. (7).

154 The same FIR filter is used at both locations which come from the same structure, and both
155 acceleration records assume to use the same sampling rate. Therefore, the estimated interstory
156 drift is determined by the difference of these displacement values from acceleration records.

157
$$\theta_i(t) = u_{i+1}(t) - u_i(t) = (\Delta t)^2 \sum_{p=-k}^k c_{k+1+p} [\bar{a}_{i+1}(t + p\Delta t) - \bar{a}_i(t + p\Delta t)] \quad (10)$$

158
$$\theta_i(t) = (\Delta t)^2 \sum_{p=-k}^k c_{k+1+p} \Delta \bar{a}_i(t + p\Delta t) \quad (11)$$

159 where $\Delta \bar{a}_i = \bar{a}_{i+1} - \bar{a}_i$ is the relative acceleration of consecutive floors.

160 The filter coefficients need to be computed only once at the beginning. In addition, the
 161 proposed method requires the multiplication of two vectors, coefficients and acceleration, to
 162 estimate interstory drifts at each time step. However, this filter is not causal, as it requires the
 163 measured relative accelerations of k points in the future. Then, this estimation is obtained with
 164 a time delay equal to half the time window length; however, this would typically lead to small
 165 lags in the estimation, and it can be implemented in near-real time.

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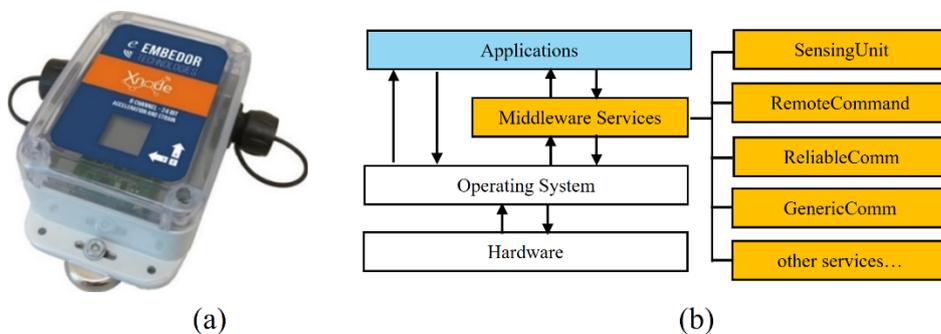
167 3. WIRELESS SMART SENSORS AND TIME SYNCHRONIZATION

168 This section presents the details about wireless smart sensors and strategies for time
 169 synchronization of the time records. The proposed method is deployed in wireless smart sensors
 170 for estimating dynamic interstory drifts, and time synchronization between different sensors is
 171 addressed by a two-point two-stage strategy proposed by the authors.

172 3.1 High-fidelity sensor platform

173 To provide high-quality measurement and high-efficiency processing for drift estimation,
 174 this study leverages a next-generation wireless smart sensor platform, the Xnode (Fig. 1)

175 (Spencer et al, 2017), because of its excellent features both in hardware and software. In
 176 particular, the Xnode employs an 8-channel, 24-bit analog-to-digital converter, which enables
 177 high-resolution data collection at a high sampling rate of up to 16kHz. In addition, the Xnode
 178 features a powerful microprocessor that operates with a dual Cortex core at frequencies up to
 179 204MHz, suitable for data-intensive on-board computation, like the interstory drift estimation.
 180 On the software side, in contrast to most commercial WSS, the Xnode is open-source, allowing
 181 users to modify and customize the software and applications. Moreover, the Xnode retains
 182 much of the successful SOA-based middleware of the Illinois Structural Health Monitoring
 183 Services Toolsuite (Rice et al, 2010) and implements it in a preemptive multitasking framework
 184 using the standard C programming language (Fu et al, 2016), which significantly facilitates the
 185 end-user development. More comprehensive performance efficacy and discussion of the Xnode
 186 can be found in the paper (Fu et al, 2018; Fu et al, 2019). Xnode smart sensor is leveraged in
 187 this study for deployment and evaluation of interstory drift estimation. Fig. 1 shows the
 188 hardware and software details of the Xnode smart sensor.



189

190

Fig. 1. Xnode smart sensor: (a) hardware platform, (b) software framework

191

192 *3.2 Two-point two-stage time synchronization for interstory drift estimation*

193 To address the challenge of time synchronization between wireless smart sensors, an
194 efficient strategy, named as two-point two-stage time synchronization, is proposed for
195 interstory drift estimation. The two stages include clock and data synchronization. This section
196 provides a brief description of the development and implementation of this strategy.

197 To prepare for the time synchronization strategy, lab tests were first conducted to quantify
198 the clock drift behavior of wireless smart sensors, Xnodes in this study. The test results reveal
199 that, without time synchronization, a drift rate achieves up to 13 $\mu\text{s/s}$, while nonlinear drift
200 behavior is negligible and sampling rate is stable (Fu et al., 2021a). Based on the observations,
201 an efficient two-point clock synchronization is developed, as shown in Fig. 2. Specifically,
202 before sensing starts, a series of beacons with global time stamps are broadcasted from the
203 gateway node to sensor nodes at an interval of 1 ms. Upon reception of beacons, each sensor
204 node records the local clock and obtain a series of corresponding clock offsets. The medium
205 value of these offsets is selected as the 1st Point of clock information before sensing starts.
206 Specifically, in this process, up to 10 beacons at 5-millisecond intervals were exchanged
207 between the gateway node and all the sensor nodes to obtain the clock offsets before sensing
208 (1st point), recorded as $\Delta t_j(i)$,

$$209 \quad \Delta t_j(i) = t_{lbj}(i) - t_{gbj}(i), \quad i \in [1,9] \quad (12)$$

210 Where $t_{gb}(i)$ is transmission time of beacon i , recorded in the gateway node, and $t_{lb}(i)$ is
211 reception time, recorded in the sensor nodes. After at least five offset values are given, the
212 median of the offset values, Δt_1 , and associated local clock, t_1 , are used as the clock
213 information pair for the current round of point synchronization. In the meantime, the current
214 clock offset is compensated, such that all the sensor nodes start sensing approximately at the
215 same designated time, T_{start} . In addition, the time to stop sensing is obtained as $T_{\text{stop}} = T_{\text{start}} +$
216 T_{sensing} , where T_{sensing} is specified by users. During sensing, the updated local clock is
217 recorded when sensing starts and stops, labeled as t_{start} and t_{stop} , respectively.

218 After sensing stops, similarly, another round of clock offset investigation is carried out,
219 and the 2nd Point of clock information obtained. Afterwards, clock drift is estimated based on
220 clock offsets obtained in the 1st Point and the 2nd Point,

$$221 \quad k = (\Delta t_2 - \Delta t_1)/(t_2 - t_1), \quad b = \Delta t_2 \quad (13)$$

222 which are used to correct the time stamp t_{lb} in each sensor node as

$$223 \quad t'_{lb} = t_{lb} - b - k(t_{lb} - t_2) \quad (14)$$

224 In addition, we can obtain the offset of start-up sensing time in Eq. (15), which will be further
225 used for the next step of data synchronization process.

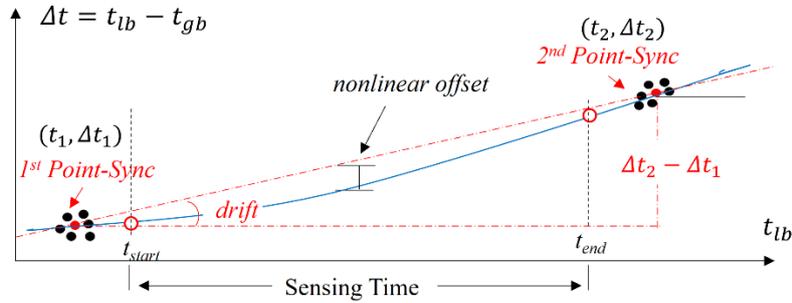
$$226 \quad dt = \frac{1}{2}(T_{\text{start}} + T_{\text{stop}} - t_{\text{start}} - t_{\text{stop}}) \quad (15)$$

227 Finally, a resampling-based approach developed by Nagayama and Spencer (2007) is
228 applied to achieve data synchronization. It will address three uncertainties for synchronization:

229 offset in start-up time, sampling rate difference among sensor nodes, and sampling rate
230 fluctuation in a single node. For the completeness of the methodology, a brief discussion about
231 this approach is conducted next. The basic idea of resampling is to achieve a signal with a factor
232 of L/M , via upsampling by L , filtering, and downsampling by M . In the process, a polyphase
233 implementation is applied to simplify the execution of an FIR filter. For a non-integer
234 downsampling factor of M to achieve a precise sampling rate, the introduction of an initial delay
235 (before upsampling and linear interpolation) is applied in the downsampling process. assuming
236 that the output data points do not necessarily correspond to the points on the upsampled signal.
237 For data synchronization, the entire sampling dataset is divided into several blocks. In each
238 block, the offset of starting time (i.e., first data point) is estimated first; and the actual sampling
239 rate is also calculated by $(t_{\text{current}} - t_{\text{last}})/N$. The timestamps after clock synchronization are
240 then used to obtain the misalignment of sample points. Finally, resampling is applied to each
241 block of data.

242 In summary, combining clock synchronization and data synchronization is proposed. This
243 strategy achieves time synchronization in an efficient way, whilst compensating the effect of
244 nonlinear clock drift. Compared with conventional methods, it is effective and accurate, which
245 is suitable for interstory drift estimation in this study. Laboratory tests were conducted to
246 evaluate the precision of the proposed approach, using three sensor nodes as leaf nodes and one
247 gateway node. The pairwise synchronization errors are collected and averaged for comparison.

248 Table 1 presents the comparison of precision between the proposed approach and the state-of-
 249 the-art solutions in the literature. The reference data is collected from the literature (Li et al.,
 250 2016), consisting of post-sensing time synchronization with linear and nonlinear regressions. It
 251 should be noted that, the difference of precision may be due to both hardware and approaches.
 252 It demonstrates that the proposed two-point two-stage effective time synchronization can
 253 achieve a time synchronization error of less than 15 μs , which is sufficiently precise for the
 254 SHM applications in this study.



255

256 **Fig. 2.** The proposed time synchronization: clock synchronization illustration

257

Table 1. Comparison of time synchronization errors

Sensing duration	Pairwise synchronization error (mean value, μs)		
	Post-sensing time synch with linear regression	Post-sensing time synch with nonlinear regression	The proposed approach
1min	18.63	18.26	7.46
10min	18.33	17.86	6.44
30min	17.63	13.27	11.48

258

259

4 NUMERICAL VALIDATIONS

260

This section considers three examples as numerical validation of the approach for linear

261

and nonlinear buildings subjected to different excitations. In real applications, nonlinear

262 behavior, such as yielding or nonlinear restoring forces, is expected in building components,
263 and therefore, the examples show the performance of the proposed method in linear buildings,
264 nonlinear hysteretic buildings, and linear building with nonlinear protection devices.

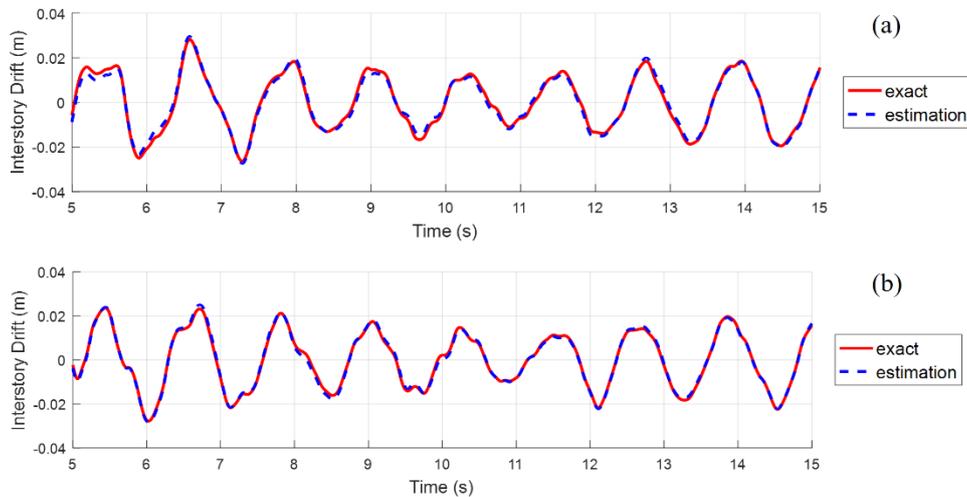
265 *4.1 Linear 9-story building*

266 In this section, a benchmark example is considered, where a 9-story linear shear building
267 is modeled subjected to ground motions (Xu et al, 2017), to numerically evaluate the accuracy
268 of the method. In particular, the mass is 505×10^3 kg for the first floor, 495×10^3 kg for the second
269 to eighth floors, and 535×10^3 kg for the roof, respectively. The stiffnesses is 600, 578, 544, 502,
270 453, 397, 332, 256, and 162 MN/m for the first to the last floor, respectively. The modal
271 damping ratio is 2% for all the modes. Regarding the excitations, we consider two different
272 ground motions, including El Centro (EC) earthquake record, and an artificial earthquake which
273 is processed by the non-stationary Kanai-Tajimi (NSKT) model (Xu et al, 2017) with the
274 properties including $\omega_g = 12$ rad/s, $\zeta_g = 0.3$, $S_0 = 0.02$ m²/s³, and $e(t) = 4(e^{-0.1t} - e^{-0.2t})$.

275 The numerical model is built and executed with a sampling rate of 1000 Hz in MATLAB
276 Simulink, and the collected response datasets are then down-sampled to the frequency of 100
277 Hz to match the popular sampling rate of wireless smart sensors (e.g., Xnodes). All interstory
278 drifts are computed for comparison purposes. All floor accelerations and base acceleration are
279 measured. To make it more realistic, datasets are added with a zero-mean Gaussian noise, where

280 standard deviation is equal to 5% of the maximum RMS acceleration. this noise magnitude has
281 been chosen based on similar previous studies (Park et al., 2018; Gomez et al., 2019).

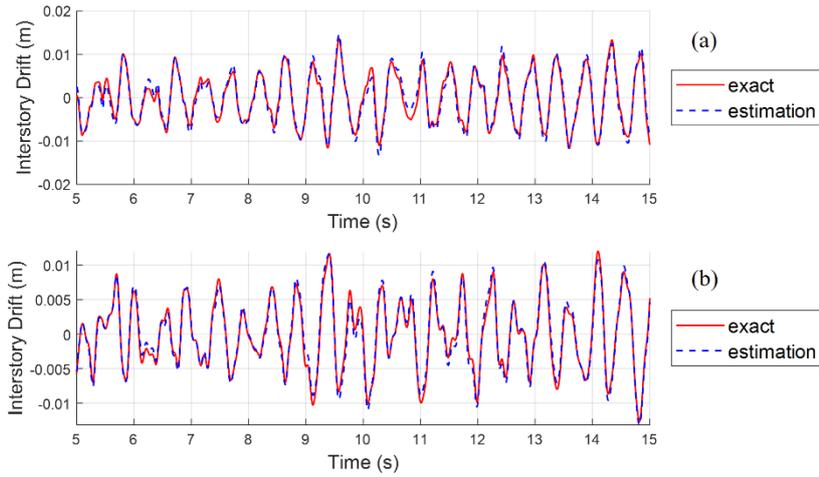
282 We use the proposed method to estimate the dynamic interstory drift in all floors using the
283 relative accelerations, using the following parameters: $f_s = 100$ Hz, $n = 4$, $f_T = 0.8$ Hz,
284 $\alpha_T = 0.99$, and $N_w = 5.223$. The total interstory drift is considered as a reference for
285 comparison, this is because in the measurements, the pseudo-static components are very small
286 and no need to extract the dynamic counterparts. Fig. 3 and 4 show the interstory drifts of floors
287 1 and 6 subjected to ground motions EC and NSKT, respectively. These figures demonstrate
288 that the obtained drifts match well with the exact values for all time steps.



289
290 **Fig. 3.** Comparison of interstory drifts under EC ground motion: (a) first story and (b) sixth story

291

292



293

294 **Fig. 4.** Comparison of interstory drifts under NSKT ground motion: (a) first story and (b) sixth story

295

296 In particular, two types of error metrics are obtained to assess the accuracy of the proposed
 297 method: the amplitude error and the RMS error. The amplitude error is given as the difference
 298 between the maximum estimated value and the maximum exact value divided by the maximum
 299 exact value. In earthquake engineering, the maximum interstory drift is one of the most
 300 important metrics (Bennett and Batrone, 1997). The magnitude of the amplitude error
 301 represents an important metric in the accuracy of the maximum interstory drift. The RMS error
 302 is defined by the root-mean-square of the difference between the estimated value and the exact
 303 value divided by the maximum exact value. Table 2 lists both types of errors under the two
 304 excitations. It demonstrates that the errors are relatively small although introducing relatively
 305 large Gaussian noise. It is also observed that, the estimations under the NSKT ground motion
 306 have larger errors than those under EC, which can be explained as the largest response occurs

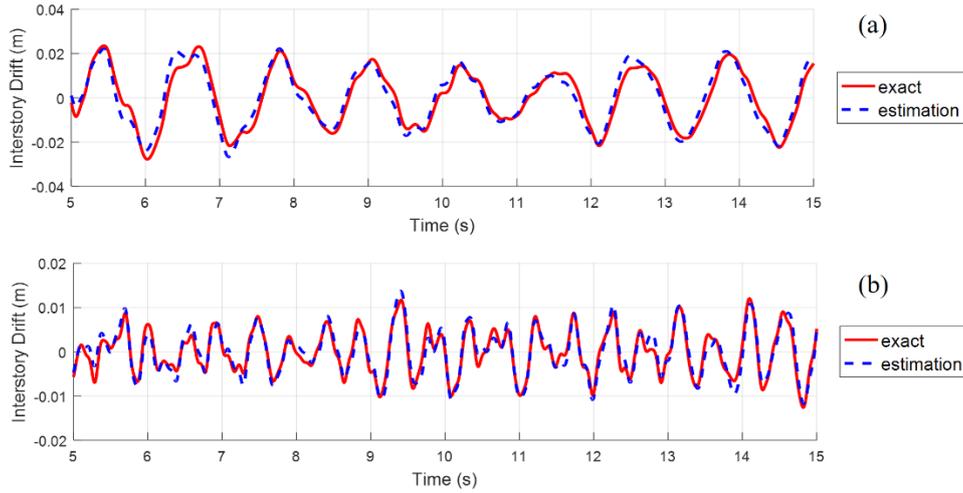
307 in one floor but all records are polluted using the Gaussian noise with the same amplitude using
 308 only the RMS of the maximum response. In the experiments, the measurement noise in
 309 acceleration measurements should be smaller than assumed values in the numerical examples.

310 **Table 2.** Estimation errors of dynamic interstory drifts in the linear 9-story building

Story	El Centro		Non-stationary Kanai-Tajimi	
	Amplitude Error (%)	RMS Error (%)	Amplitude Error (%)	RMS Error (%)
1	4.89	3.42	1.26	6.17
2	2.03	3.75	6.30	6.98
3	4.77	4.30	2.36	7.99
4	4.73	3.47	1.37	9.59
5	1.43	4.29	1.60	8.37
6	1.17	3.31	6.08	6.94
7	1.02	2.83	7.04	4.46
8	5.38	2.89	2.82	4.30
9	0.79	2.87	5.97	3.22

311
 312 As discussed in Section 1, measurements from different wireless smart sensors have time
 313 lags between them. Now, to assess the effect of this issue in the estimation, a time lag of 10
 314 milliseconds is introduced in the acceleration record of the sixth floor. Clearly, this change will
 315 greatly affect the interstory drift of the sixth and seventh floor.

316 Fig. 5 shows the comparison of the interstory drift for 6th floor for both ground motions.
 317 The estimated and exact values do not agree as well as before. The time lag in the data harms
 318 the accuracy of the method. Smaller time lags need to be assured to achieve good results as
 319 before.



320

321 **Fig. 5.** Comparison of interstory drifts for sixth story with a time lag of 10 ms: (a) EC and (b) NSKT

322 ground motions

323

324 Table 3 shows the amplitude and RMS error for both excitations. The amplitude errors are

325 slightly larger but do not change much from the previous case. However, the RMS error is

326 considerably amplified because the time lag not only affects the amplitude but also introduces

327 a lag between the estimation and exact measurements. In sum, a strategy to limit time lags in

328 data from different wireless sensors is needed to achieve good accuracy in the drift estimation.

329

330 **Table 3.** Estimation errors with a time lag of 10 milliseconds at sixth floor acceleration

Story	El Centro		Non-stationary Kanai-Tajimi	
	Amplitude Error (%)	RMS Error (%)	Amplitude Error (%)	RMS Error (%)
6	1.64	13.13	5.53	15.44
7	10.64	12.14	11.03	8.09

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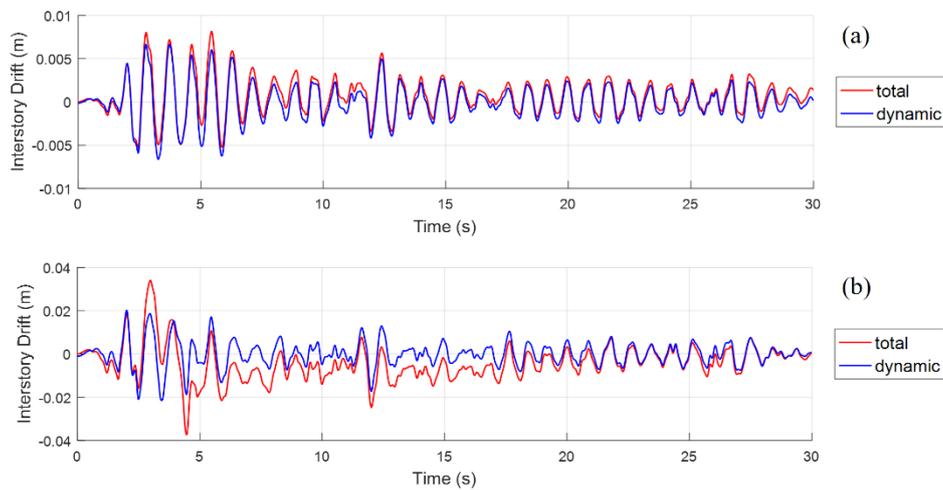
332 4.2 Nonlinear 3-story building

333 A 3-story nonlinear hysteretic shear building is applied in this section to test the accuracy
334 of the method, subjected to ground motions (Xu et al, 2017). The mass of each floor is 6000
335 kg. The linear stiffnesses for the first to the last floor are 2.178, 1.772, and 1.2969 MN/m,
336 respectively. The damping ratios are assumed to be 1.6%, 1.7%, and 2.7% for each mode. Each
337 floor is considered as an elastoplastic element with smooth transition using Bouc-Wen
338 hysteretic behavior; typical parameters are assumed for each floor:
339 $\gamma = \beta = 0.5$, $A = 1$, $n = 1$, $\alpha = 0.04$, $d_y = 0.01$. Two ground motions are considered: El Centro
340 (EC) earthquake record scaled to 20% and with no scaling; for larger amplitudes of the record,
341 the response should have a larger nonlinear component. The excitation, Simulink simulation,
342 and data post-processing are the same as those set in Section 4.1.

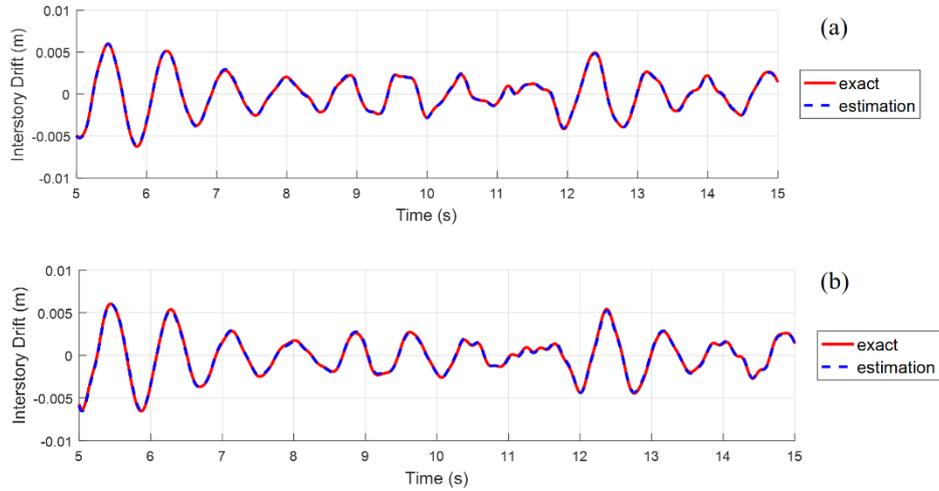
343 In this case, the elastoplastic behavior implies residual deformation due to yielding in the
344 system and this phenomenon introduces pseudo-static displacements; this phenomenon
345 increases as the amplitude of the excitation is increased. Therefore, the dynamic interstory drift
346 is extracted from the measured total interstory drift to provide a comparison with the estimated
347 interstory drift. Fig. 6 shows the total and dynamic interstory drifts of the first story for both
348 excitations. It should be clarified that, for strong earthquakes, the response is expected to consist
349 of large residual deformations; however, the dynamic interstory drift is still considered useful
350 for rapid condition assessment of buildings (Fu et al, 2021b). Other types of nonlinear behavior

351 such as components with nonlinear restoring forces do not introduce residual deformations and
352 the total interstory drift can be recovered with the proposed method; this case is presented in
353 the next subsection.

354 The proposed method is applied to estimate dynamic interstory drifts using measured
355 accelerations, and it considers the following parameters: $f_s = 100$ Hz, $n = 4$, $f_T = 1.2$ Hz,
356 $\alpha_T = 0.99$, and $N_w = 5.223$. Fig. 7 and 8 show the comparison of the dynamic interstory drift
357 for floors 1 and 2 and for both excitations. As these figures show, the estimated interstory drift
358 agrees well with exact dynamic interstory drift for all time steps. Table 4 shows the amplitude
359 and RMS errors of interstory drift estimation for both excitations. As can be seen in the figure,
360 the errors between the estimated and the exact dynamic interstory drifts are relatively small.



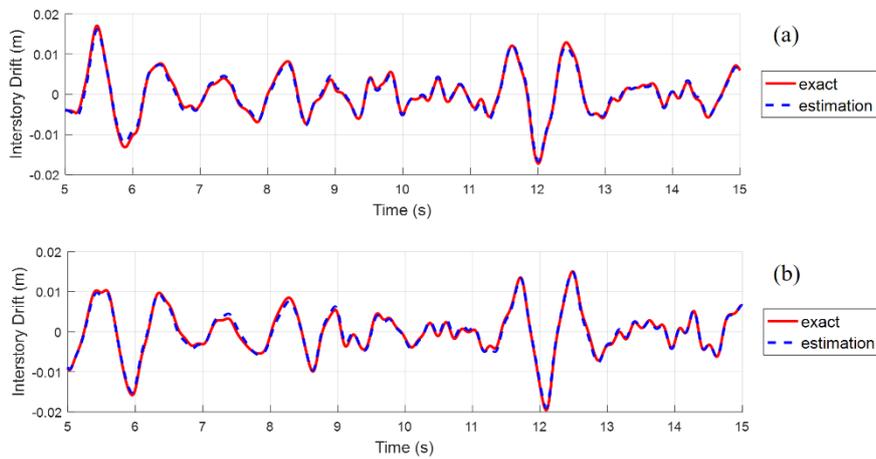
361
362 **Fig. 6.** Total and dynamic interstory drifts of the first story with EC (a) scaled to 20% and (b) unscaled



363

364

Fig. 7. Comparison of interstory drifts with EC scaled to 20%: (a) first and (b) second story



365

366

Fig. 8. Comparison of interstory drifts with EC unscaled: (a) first and (b) second story

367

Table 4. Estimation errors of dynamic interstory drifts in the nonlinear 3-story building

Story	El Centro 20%		El Centro 100%	
	Amplitude Error (%)	RMS Error (%)	Amplitude Error (%)	RMS Error (%)
1	2.40	1.31	3.79	2.56
2	0.05	1.44	3.01	2.11
3	1.12	1.74	0.90	3.12

368

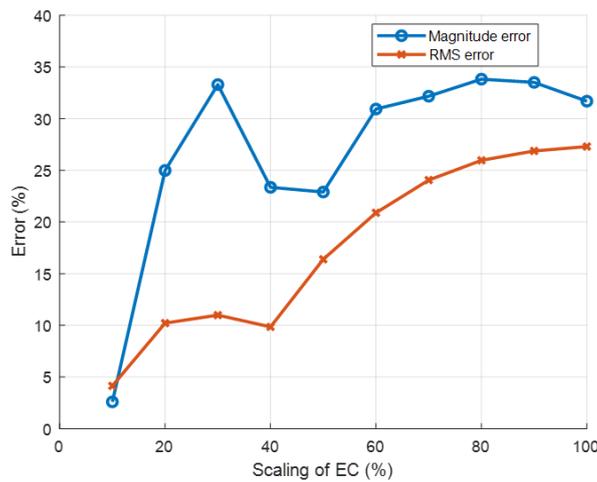
369

To study the effect of nonlinear effects on the accuracy of the proposed method for total

370

interstory drifts, the analysis was performed for different scalings of El Centro. The magnitude

371 error and RMS error for the different scaling are shown in Fig. 9. Both types of errors increase
372 as the nonlinear behavior increases. Moreover, unless large nonlinear plastic behavior, which
373 achieves residual deformations, is expected to occur, the error in the estimated interstory drift
374 with respect to the total interstory drift is reasonable.



375
376 **Fig. 9.** Comparison of errors for the total interstory drifts of the first floor for different levels of EC

377
378 *4.3 Linear 2-story building with a nonlinear energy sink*

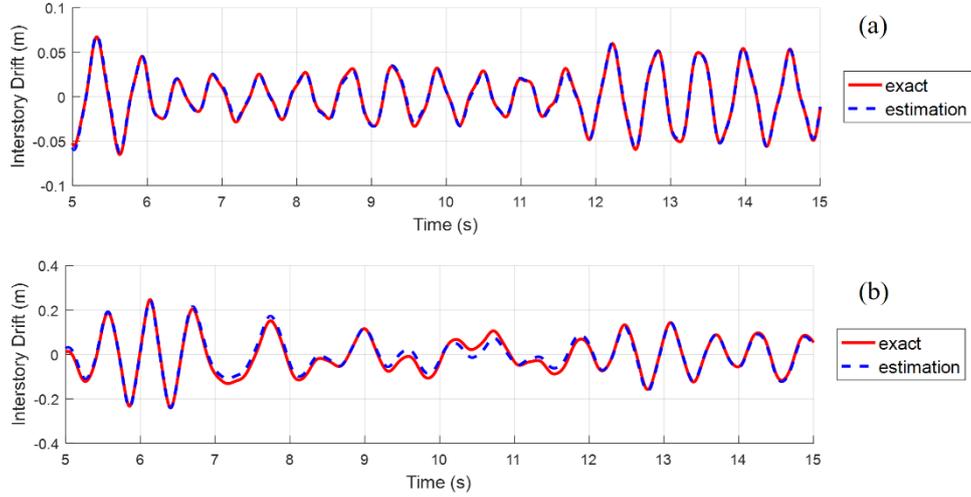
379 Buildings with seismic protective devices such as nonlinear dampers or isolators typically
380 have a nonlinear behavior with limited pseudo-static residual deformations, and interstory drift
381 measurement of these nonlinear systems is of interest as well. This example is representative
382 of buildings with non-linear protection devices. A 2-story linear shear building with a nonlinear
383 energy sink (NES) on the roof subjected to ground motions (Gomez et al, 2021) is considered

384 as an example of these systems. In this case, the device stroke, which is the relative
385 displacement of the floor and the device, is also a response of interest.

386 The masses of the floors are 24.3 and 24.2 kg. The linear stiffnesses are 6820 and 8220
387 N/m. The damping ratios are assumed to be 0.1% for the two modes of the uncontrolled
388 structure. The NES in the roof is a Duffing oscillator without a linear term with mass equal to
389 6.81% of the total mass, linear damping force with constant 3.57 N-s/m, and the coefficient
390 $\alpha_N = 63.97 \text{ m}^{-3}$ [34]. Two ground motions are considered: EC earthquake record and NSKT
391 model with the following properties: $\omega_g = 20.3 \text{ rad/s}$, $\zeta_g = 0.32$, $S_0 = 0.026 \text{ m}^2/\text{s}^3$, and
392 $e(t) = 4(e^{-0.1t} - e^{-0.2t})$ (Gomez et al, 2021). Due to the essential nonlinearity in the NES, the
393 system always has a nonlinear behavior. The excitation, Simulink simulation, and data post-
394 processing are the same with those set in Section 4.1.

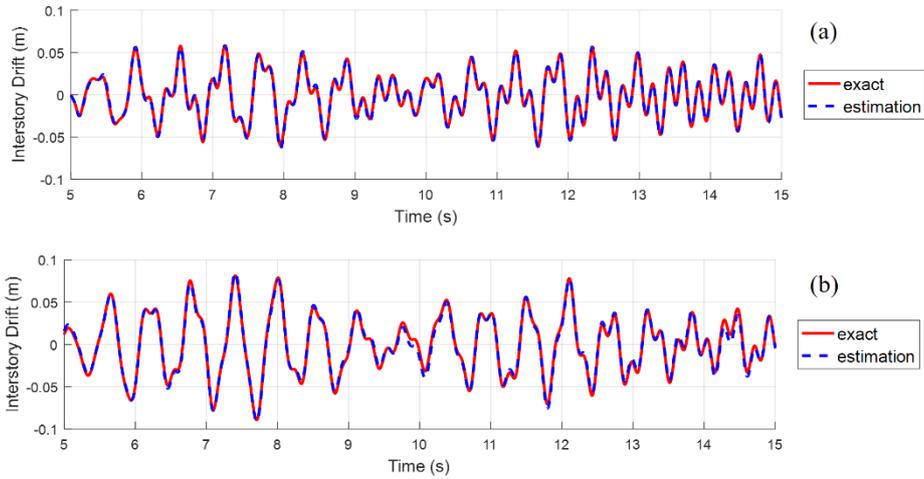
395 The proposed method is applied to estimate dynamic interstory drifts from measured
396 accelerations with the following parameters: $f_s = 100 \text{ Hz}$, $n = 4$, $f_T = 1.0 \text{ Hz}$, $\alpha_T = 0.99$,
397 and $N_w = 5.223$. The total interstory drift is considered as a reference for comparison, this is
398 because in the measurements, the pseudo-static components are very small and no need to
399 extract the dynamic counterparts. Fig. 10 and 11 show the comparison of the interstory drift for
400 the first floor and device stroke subjected to ground motions EC and NSKT, respectively. It can
401 be demonstrated that the estimated drifts agree well with the exact values. Table 5 shows both

402 types of errors for both excitations. It demonstrates that the errors are relatively small although
 403 relatively large Gaussian noise are introduced.



404

405 **Fig. 10.** Comparison with EC for (a) interstory drift of first floor and (b) device stroke



406

407 **Fig. 11.** Comparison with NSKT for (a) interstory drift of first floor and (b) device stroke

408 **Table 5.** Estimation errors of the dynamic interstory drifts in the linear 2-story building

Story	El Centro		Non-stationary Kanai-Tajimi	
	Amplitude Error (%)	RMS Error (%)	Amplitude Error (%)	RMS Error (%)
1	2.43	2.34	4.86	2.25
2	0.82	5.71	1.87	3.79
NES	1.78	5.87	1.57	3.51

409

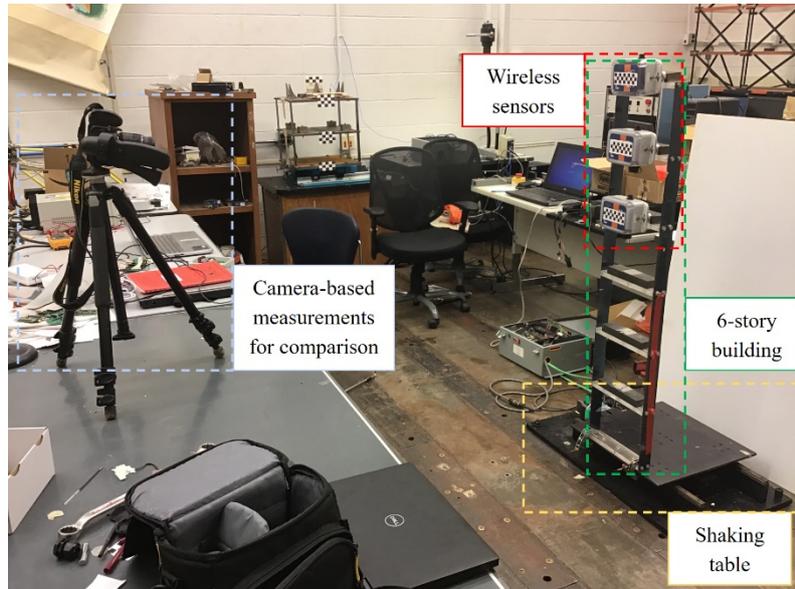
5. LABORATORY VALIDATION

410 This section presents a laboratory validation of the proposed method. First, the
411 experimental setup is described. Then, the results and their discussion are presented.

412 *5.1 Experimental setup*

413 A 6-story planar steel frame was excited using a uniaxial shaking table; based on traditional
414 system identification methods, the first frequencies of the frame were 1.63 Hz and 5.13 Hz, and
415 first modal damping ratios are 3.9% and 1.9%. The absolute accelerations were measured using
416 3 smart wireless sensors Xnodes at floors 4, 5, and 6; the data acquisition was set to 100 Hz.
417 Vision-based measurements were also obtained to extract displacements and used as a reference
418 for comparison. Specifically, a checkerboard pattern was attached to each sensor, visible to the
419 camera and acting as target for tracking. Due to limited size of the pattern (smaller than 4-by-
420 4), the MATLAB toolbox which was used to detect and track the checkerboard pattern was not
421 applicable (Calibrator, 2019). Thus, a simple pattern matching using 2-dimensional cross-
422 correlation was used on each frame to track the displacement of each floor in pixel unit. Then
423 the measurement was converted to mm given the size of each square of the checkerboard pattern
424 was 20 mm x 20 mm. Nikon D3300 camera with the lens of 18-55mm was used, and data
425 acquisition was set to 60 frames-per-second for video recording. Fig. 12 shows the test setup.

426 Three ground motion records with different dynamic properties were considered as the
427 motion of the shaking table to excite the structure: El Centro 1940, Northridge 1994, and Kobe
428 1995.



429

430

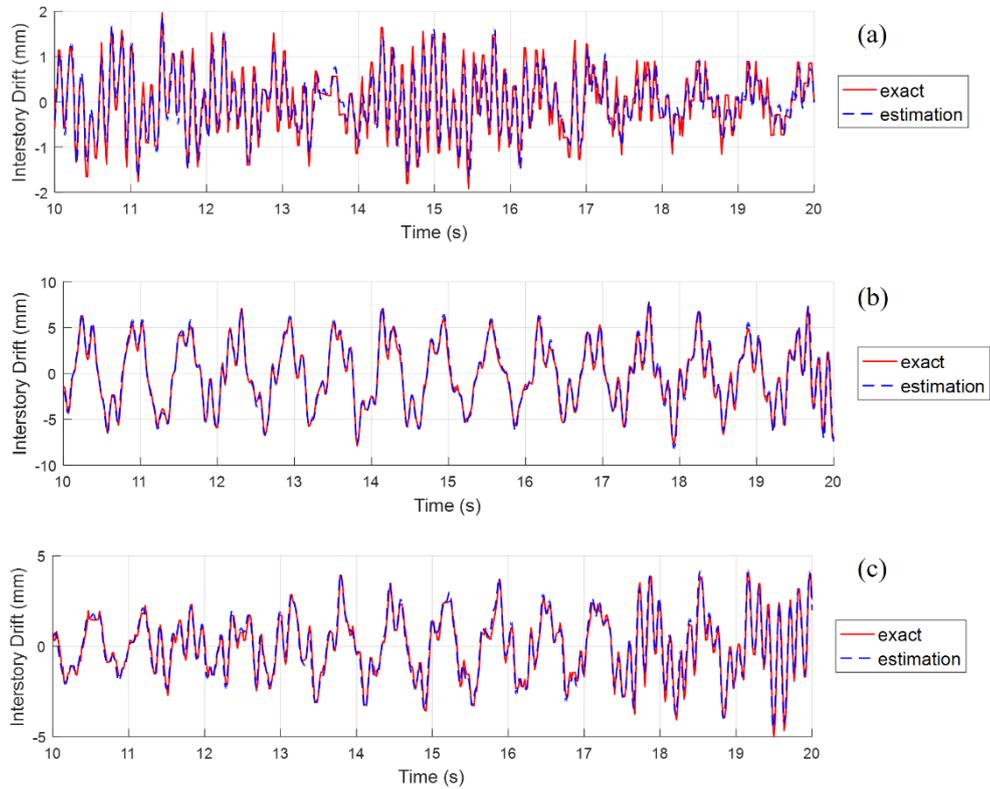
Fig. 12. Experimental setup

431

432 *5.2 Results and Discussion*

433 The proposed dynamic interstory drift estimation method is applied to the measured
434 accelerations, and it considers the following parameters: $f_s = 100$ Hz, $n = 4$, $f_T = 1.2$ Hz,
435 $\alpha_T = 0.99$, and $N_w = 5.223$. The total interstory drift is considered as a reference for
436 comparison, this is because in the measurements, the pseudo-static components are very small
437 and no need to extract the dynamic counterparts. Fig. 13a-c show the comparison of a time-
438 window of the dynamic interstory drift estimation for floor 5 against the camera-based

439 measurement subjected to all ground motions. It can be demonstrated that the estimated values
440 and the exact values match well for all time steps.



441
442 **Fig. 13.** Comparison of interstory drifts for 5th story (a) El Centro, (b) Northridge, and (c) Kobe
443 earthquakes

444
445 The results of the experimental validation for all cases indicate that the proposed interstory
446 drift estimation, such that the maximum magnitude error is smaller than 5.5% for all cases.
447 Table 6 shows the errors in the proposed method compared to camera-based measurements. It
448 is worth noting that the comparison is done against the total interstory drift because the pseudo-
449 static component is negligible. It is concluded that the method is adequate for interstory drift

450 estimation in both amplitude and phase. Currently, many building structures include nonlinear
 451 protection devices. As the nonlinearities occur at the discrete locations, where these devices are
 452 located, the proposed approach also works well for estimating total interstory drifts in these
 453 structures.

454 **Table 6.** Estimation errors of the dynamic interstory drifts for laboratory validation

Excitation	Story	Amplitude Error (%)
El Centro	5	4.46
	6	5.10
Northridge	5	5.42
	6	4.74
Kobe	5	5.31
	6	4.24

455

456

457

6. CONCLUSIONS

458 This paper proposes a new method to estimate dynamic interstory drifts in buildings using
 459 measured accelerations from smart wireless sensors. The method uses an effective FIR filter
 460 that is suited to better suppress low-frequency noise with small ripple in the passband, based on
 461 a minimization problem with Tikhonov regularization. This method then makes use of the
 462 difference of the measured acceleration from different floors, which requires the measurements
 463 to be time-synchronized. Dynamic interstory drift estimation using acceleration measurements
 464 has the potential to be implemented by leveraging WSS. But time synchronization must be
 465 addressed first between WSS nodes. A two-point two-stage method to efficiently perform time
 466 synchronization of multiple WSS is presented to reduce the errors to exceptionally small values.

467 The proposed method was demonstrated and evaluated via numerical simulations of both linear
468 and nonlinear hysteretic buildings subjected to ground motions, and subsequently demonstrated
469 in laboratory tests of a small-scale steel frame subjected to different earthquake records. Both
470 numerical and lab test results demonstrate that the proposed method provides a very accurate
471 estimation of the dynamic interstory drifts of buildings.

472 Future work in this topic will consist in improving the filter by making the window of the
473 filter shorter, such that the lag between the measurement and prediction becomes smaller and
474 implementation is more efficient as fewer arithmetic operations are needed in the WSS.
475 Additionally, the filter will be studied to include an estimation of the pseudo-static interstory
476 drifts in nonlinear structures. Experiments on large-scale buildings will be conducted.

477 DATA AVAILABILITY STATEMENT

478 Some data, models, and code that support the findings of this study are available from the
479 corresponding author upon reasonable request.

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